

Adaptation and the effort needed to adapt

Sergio Bittanti, Marco C. Campi, and Maria Prandini

Abstract—Tuning a system to an operating environment calls for experimentation, and a question that arises naturally is: how many experiments are needed to come up with a system meeting certain performance specifications?

This paper is an attempt to answer this fundamental question at a somehow general level. We shall here refer to a set-up where adaptation is done according to a worst-case approach, but many facts established are central to adaptation in general.

I. INTRODUCTION

Given a system \mathcal{S} , consider the problem of designing a device \mathcal{D} that achieves some desired behavior when interacting with \mathcal{S} . The specification of the ‘desired behavior’ depends on the intended use of the device, and is usually expressed by saying that some signal $s_{\mathcal{D}}$ generated by the interaction of \mathcal{S} and \mathcal{D} should behave in a proper and desired way. To emphasize that the generated signal $s_{\mathcal{D}}$ also depends on the specific condition of operation, in the sequel we shall use $\omega \in \Omega$ to indicate the operating condition and write $s_{\mathcal{D}}(\omega)$ for the generated signal, see Figure 1.

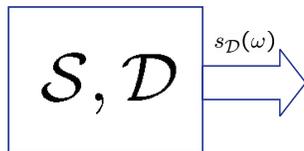


Fig. 1. Signal $s_{\mathcal{D}}(\omega)$ is generated by the interaction of device \mathcal{D} and system \mathcal{S} .

This idea is made more concrete through examples.

Example 1 (simulator): Suppose that the device should act as a simulator of the system when the system input u takes on value in a given class of signals U . In this case, $\omega = u$ and the desired behavior for the device can be expressed in terms of the multidimensional signal $s_{\mathcal{D}}(u) = (y(u), y_{\mathcal{D}}(u))$, where $y_{\mathcal{D}}(u)$ and $y(u)$ are the outputs of the device and the output of the system fed by the same input signal $u \in U$ (see Figure 2). The signal $s_{\mathcal{D}}(u)$ should satisfy $y_{\mathcal{D}}(u) \simeq y(u)$, for every operating condition of interest, that is for every $u \in U$. \square

Example 2 (disturbance compensator): Suppose that the output of system \mathcal{S} is affected by some additive disturbance

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S. Bittanti and M. Prandini are with the Dipartimento di Elettronica e Informazione - Politecnico di Milano, piazza Leonardo da Vinci 32, 20133 Milano, Italy {bittanti, prandini}@elet.polimi.it

M.C. Campi is with the Dipartimento di Elettronica per l’Automazione - Università di Brescia, via Branze 38, 25123 Brescia, Italy marco.campi@ing.unibs.it

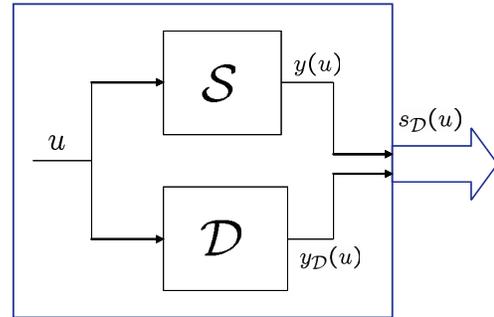


Fig. 2. Device \mathcal{D} is as a simulator of system \mathcal{S}

and the device \mathcal{D} is introduced for compensating the disturbance according to a feedforward scheme as in Figure 3. In this case the operating condition is defined by the disturbance realization d , so that $\omega = d$. If we denote by $y_{\mathcal{D}}(d)$ the controlled output of the system when the disturbance realization is d , then the desired behavior can be expressed in terms of the signal $s_{\mathcal{D}}(d) = y_{\mathcal{D}}(d)$ and $s_{\mathcal{D}}(d)$ should be small for every d in some set D describing the disturbances. \square

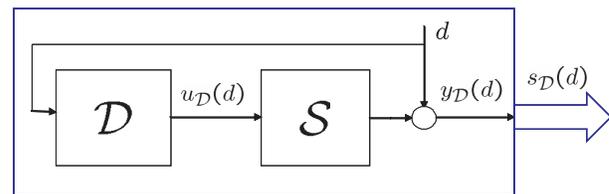


Fig. 3. Device \mathcal{D} acting as a disturbance compensator for system \mathcal{S}

Devising a suitable device \mathcal{D} for a system \mathcal{S} requires knowledge of some sort on \mathcal{S} . Most literature in science and engineering relies on a model-based approach, namely it is assumed that a mathematical model for \mathcal{S} is a-priori available. Alternatively, the knowledge on \mathcal{S} can be accrued through experimentation. This latter approach, considered herein, is referred to as ‘adaptive design’, [1], [2], [3], [4], [5], since the problem is to adapt \mathcal{D} on the basis of experiments in the face of the lack of a-priori knowledge on system \mathcal{S} .

In adaptive design, one fundamental question to ask is:

How extensively do we need to experiment in order to come up with a device meeting certain performance requirements?

This fundamental –and yet largely unanswered– question is

the theme this contribution is centered around.

In this paper, a *worst-case perspective* with respect to the possible operating conditions is adopted, and we provide an answer to the above question in this specific set-up. For one answer, many more are the answers that this contribution is incapable to provide, which will also be enlightened along our way.

II. WORST-CASE APPROACH TO ADAPTATION

Worst-case performance

Suppose that the performance of the device \mathcal{D} operating in condition ω is quantified by a cost $c(s_{\mathcal{D}}(\omega))$. Then, the worst-case performance achieved by \mathcal{D} over the set Ω of operating conditions is

$$\max_{\omega \in \Omega} c(s_{\mathcal{D}}(\omega)),$$

and, correspondingly, one wants to design

$$\mathcal{D}^* = \arg \min_{\mathcal{D}} \max_{\omega \in \Omega} c(s_{\mathcal{D}}(\omega)). \quad (1)$$

c^* denotes the worst-case performance of device \mathcal{D}^* , that is $c^* = \max_{\omega \in \Omega} c(s_{\mathcal{D}^*}(\omega))$.

In e.g. the simulator Example 1, $\omega = u$ and one can take $c(s_{\mathcal{D}}(u)) = \|y(u) - y_{\mathcal{D}}(u)\|_2$, the 2-norm of the error signal $y(u) - y_{\mathcal{D}}(u)$. c^* can then be interpreted as an upper bound to the largest 2-norm discrepancy between the system behavior and the behavior of the simulator \mathcal{D}^* in the same operating condition:

$$\|y(u) - y_{\mathcal{D}^*}(u)\|_2 \leq c^*, \quad \forall u \in U.$$

In the disturbance compensator Example 2, a sensible cost is the 2-norm $c(s_{\mathcal{D}}(d)) = \|y_{\mathcal{D}}(d)\|_2$. Then, the best disturbance compensator \mathcal{D}^* satisfies:

$$\|y_{\mathcal{D}^*}(d)\|_2 \leq c^*, \quad \forall d \in D.$$

In many cases, the device \mathcal{D} is parameterized by a vector $\gamma \in \mathbb{R}^k$, in which case we write \mathcal{D}_{γ} to indicate device \mathcal{D} with parameter γ , and hence designing a device corresponds to selecting a value for γ . Then, with the shorthand

$$J_{\gamma}(\omega) := c(s_{\mathcal{D}_{\gamma}}(\omega)),$$

the min-max optimization problem (1) can be rewritten as the following robust optimization program with $k + 1$ optimization variables:

$$\text{RP: } \min_{\gamma, c \in \mathbb{R}^{k+1}} c \quad \text{subject to:} \quad (2)$$

$$J_{\gamma}(\omega) \leq c, \quad \forall \omega \in \Omega.$$

Note that, given a γ , the slack variable c represents an upper bound for cost $J_{\gamma}(\cdot)$ achieved by the device with parameter γ . By solving (2) we seek that γ^* that corresponds to the smallest upper bound c^* .

Adaptive design

In model-based design, the cost $J_{\gamma}(\omega)$ can be evaluated based on the model, and then γ^* is found by solving the robust optimization program (2). Instead, when system \mathcal{S} is unknown, or only partially known, the cost $J_{\gamma}(\omega)$ cannot be explicitly computed so that the constraints in (2) are not known. However, one can conceive of evaluating the constraints experimentally. What exactly this means is discussed in the sequel.

Each constraint is associated with an operating condition $\omega \in \Omega$. Evaluating experimentally a constraint in a specific condition, say condition $\hat{\omega} \in \Omega$, amounts to experimentally determine the domain of feasibility in the (γ, c) 's space where the constraint $J_{\gamma}(\hat{\omega}) \leq c$ holds. To this purpose, a sequence of experiments should be run in the $\hat{\omega}$ condition, each of which performed with a different device \mathcal{D}_{γ} , $\gamma \in \mathbb{R}^k$, and the corresponding performance $J_{\gamma}(\hat{\omega})$ should be evaluated from the measured $s_{\mathcal{D}_{\gamma}}(\hat{\omega})$. An obvious objection to this way of proceeding is that it would require in principle to test the performance achieved by each and every device \mathcal{D}_{γ} . It is an interesting fact that, in many situations, such overwhelming experimental effort can be avoided, and just one single experiment is enough for the purpose of computing $J_{\gamma}(\hat{\omega})$.

Take e.g. the simulator Example 1. In this example, if \hat{u} is injected into \mathcal{S} , signal $\hat{y} = \mathcal{S}[\hat{u}]$ can be collected, along with signal \hat{u} itself. Based on this single experiment, one can then compute $y(\hat{u}) - y_{\mathcal{D}_{\gamma}}(\hat{u}) = \hat{y} - \mathcal{D}_{\gamma}[\hat{u}]$ for all γ 's, where $\mathcal{D}_{\gamma}[\hat{u}]$ is obtained by processing \hat{u} with \mathcal{D}_{γ} , an operation that can be executed as an off-line post-processing of signal \hat{u} . After $y(\hat{u}) - y_{\mathcal{D}_{\gamma}}(\hat{u})$ has been computed, the constraint $\|y(\hat{u}) - y_{\mathcal{D}_{\gamma}}(\hat{u})\|_2 = J_{\gamma}(\hat{u}) \leq c$ is evaluated.

The same conclusion that one experiment is enough can also be drawn for Example 2 whenever both the system and the device are linear. Indeed, swapping the order of \mathcal{S} and \mathcal{D}_{γ} , we have:

$$y_{\mathcal{D}_{\gamma}}(d) = \mathcal{S}[\mathcal{D}_{\gamma}[d]] + d = \mathcal{D}_{\gamma}[\mathcal{S}[d]] + d. \quad (3)$$

If we run an experiment in which disturbance \hat{d} is measured and this disturbance is also injected as input to the system (i.e. \mathcal{D} is set to 1 during experimentation in the scheme of Figure 3), from the measured system output $\hat{y} = \mathcal{S}[\hat{d}] + \hat{d}$ and from \hat{d} itself we can then determine

$$\begin{aligned} y_{\mathcal{D}_{\gamma}}(\hat{d}) &= \mathcal{D}_{\gamma}[\mathcal{S}[\hat{d}]] + \hat{d} \quad (\text{using (3)}) \\ &= \mathcal{D}_{\gamma}[\hat{y} - \hat{d}] + \hat{d}, \end{aligned}$$

where computation of $\mathcal{D}_{\gamma}[\hat{y} - \hat{d}]$ is executed off-line similarly to the simulator example. By computing $\|y_{\mathcal{D}_{\gamma}}(\hat{d})\|_2 = J_{\gamma}(\hat{d})$ constraint $J_{\gamma}(\hat{d}) \leq c$ is then evaluated.

In the sequel we shall assume that one single experiment in condition $\hat{\omega}$ suffices to determine constraint $J_{\gamma}(\hat{\omega}) \leq c$. This assumption is not fulfilled in all applications of the adaptive scheme, and further discussion on this point is provided in Section V.

Remark 1: The reader may have noticed that lack of knowledge, for which adaptation is required, can enter the problem in different ways. In Example 1, it was system \mathcal{S} to be unknown. In the disturbance compensator Example 2, again uncertainty stayed with the system \mathcal{S} , but even the set D for d could be unknown.

The seemingly different nature of the uncertainty in \mathcal{S} and in D can be leveled off by adopting a more abstract behavioral perspective, [6], where the system is just seen as a set of behaviors, i.e. of possible realizations of system signals. In such framework, uncertainty simply corresponds to say that the set of behaviors defining the system is not a-priori known. \square

We are now facing the central issue this contribution is centered around, that is: an exact solution of the robust optimization program (2) requires to consider as many experiments as the number of elements in Ω , normally an *infinite* number. The impossibility to carry out this task suggests introducing approximate schemes where only a *finite* number of ω 's, that is a *finite* number of experiments, are considered. Thus, we can at this point more precisely spell out the question we posed at the end of Section I, and ask:

How many experiments do we need to perform in order to come up with a design that approximates the solution \mathcal{D}^ of (2) to a desired level of accuracy?*

III. THE EXPERIMENTAL EFFORT NEEDED FOR ADAPTATION

The fact that one concentrates on a finite number of operating conditions only may appear naive. The interesting fact is that this way of proceeding can be cast within a solid mathematical theory providing us with guarantees on the level of accuracy obtained.

Fix an integer N , and let $\omega^{(1)}, \omega^{(2)}, \dots, \omega^{(N)} \in \Omega$ be the operating conditions of N experiments run on the system to evaluate the N corresponding constraints for the robust program (2). The robust optimization problem restricted to the N experienced scenarios $\omega^{(i)}$, $i = 1, 2, \dots, N$, reduces to the following finite optimization problem referred in the sequel to as ‘scenario program’:

$$\text{SP}_N : \min_{\gamma, c \in \mathbb{R}^{k+1}} c \quad \text{subject to:} \quad (4)$$

$$J_\gamma(\omega^{(i)}) \leq c, \quad i = 1, 2, \dots, N.$$

Being finite, this problem has a level of solvability much higher than the hard RP problem in (2) carrying infinitely many constraints. As for the selection of the scenarios $\omega^{(i)}$, $i = 1, 2, \dots, N$, we suppose that they are extracted from set Ω according to some probability distribution P that reflects the likelihood of the different ω situations. This is naturally the case in the disturbance compensator Example 2 where the disturbances are obtained in experimental trials. If the scenarios are instead selected by the designer of the experiment, like u in Example 1, probability P

is artificially introduced to describe the likelihood of the different operating conditions.

Let (γ_N^*, c_N^*) be the solution of SP_N . c_N^* quantifies the performance of the device with parameter γ_N^* over the extracted operating conditions $\omega^{(1)}, \omega^{(2)}, \dots, \omega^{(N)}$. Moreover, we clearly have $c_N^* \leq c^*$, the optimal cost with all the constraints in place, and, for the extracted scenarios, we have designed a very effective device, even ‘better than the best’, in the sense that it outperforms device \mathcal{D}^* . We cannot be satisfied with this sole result, however, since, due to the limited number of scenarios, there is no performance guarantee for the much larger multitude of possible operating conditions, all those that have not been seen when performing the design of γ_N^* . Hence, the following question arises naturally: what can we claim regarding the performance of the designed device for all other operating conditions $\omega \in \Omega$, those that were not experienced while doing the design according to SP_N in (4)? Answering this question is necessary to provide accuracy guarantees and to pose the method on solid grounds. And, indeed, our goal is to quantify the required number N of experiments to ensure certain performance levels even on unseen operating conditions.

The posed question is of the ‘generalization type’ in a learning-theoretic sense: we want to know how the solution (γ_N^*, c_N^*) generalizes from experienced operating conditions to unexperienced ones. For ease of explanation, we shall henceforth concentrate on robust optimization problems of *convex-type*, since this case can be handled in the light of a powerful theory that has recently appeared in the literature of robust optimization, [7], [8], [9]. The non-convex case can be dealt with along a more complicated approach and is not discussed herein.

The following result is taken from [9].

KEY RESULT: Select a ‘violation parameter’ $\epsilon \in (0, 1)$ and a ‘confidence parameter’ $\beta \in (0, 1)$. If N satisfies

$$\sum_{i=0}^k \binom{N}{i} \epsilon^i (1-\epsilon)^{N-i} \leq \beta \quad (5)$$

(recall that $k+1$ is the number of optimization variables), then, with probability no smaller than $1-\beta$, the solution (γ_N^*, c_N^*) to (4) satisfies all constraints of problem (2) with the exception of those corresponding to a set of operating conditions whose probability is at most ϵ .

Let us try to explain in detail the meaning of this result. If we neglect for a moment the part associated with the confidence parameter β , then, the result simply says that, by extracting a number N of operating conditions as given by (5) and running the corresponding N experiments to evaluate the constraints appearing in (4), the solution (γ_N^*, c_N^*) to (4) violates the constraints corresponding to other, unexperienced, operating conditions with a probability that does not exceed a *user-chosen* level ϵ . This means that

the so-determined c_N^* provides an upper bound for the cost $J_{\gamma_N^*}(\omega)$ valid for every operating condition $\omega \in \Omega$ with the exclusion of at most an ϵ -probability set. What may be surprising is that this result holds irrespective of the functional dependence of $J_\gamma(\omega)$ on ω .

As for the probability $1 - \beta$, one should note that (γ_N^*, c_N^*) is a random quantity because it depends on the randomly extracted operating conditions $\omega^{(1)}, \omega^{(2)}, \dots, \omega^{(N)}$. It may happen that these conditions are not representative enough (one could even extract N times the same operating condition!). In this case no generalization is expected, and the fraction of operating conditions violated by (γ_N^*, c_N^*) will be larger than ϵ . Parameter β controls the probability of extracting unrepresentative operating conditions, and the final result that (γ_N^*, c_N^*) violates at most an ϵ -fraction of operating conditions holds with probability $1 - \beta$. One important practical fact is that, due to the structure of the equation in (5), β can be set to be so small (say $\beta = 10^{-6}$) that it is virtually zero for any practical purpose, and this does not lead to a significant increase in the value of N (see also the numerical example in Section IV).

For the reader's convenience, the discussion in this section is summarized in a recipe for a practical implementation of the overall adaptive design scheme.

PRACTICAL RESULT: *Select a violation parameter $\epsilon \in (0, 1)$, let $\beta = 10^{-6}$, and compute the smallest integer N satisfying (5). Run N random experiments and compute the corresponding N constraints for problem (4).*

Then, the solution γ_N^ of (4) achieves performance c_N^* on all operating conditions but an ϵ fraction of them, and, moreover, $c_N^* \leq c^*$.*

Before closing the section, the following final remark is worth making in the light of equation (5):

The number of experiments N that are needed to adapt the device does not depend on the system complexity; it instead only depends on the complexity of the device \mathcal{D}_γ through the size k of its parametrization γ .

Thus reality can be any complex and still we can evaluate the experimental effort by only looking at the device being designed.

IV. A NUMERICAL EXAMPLE

We consider the problem of inverting the nonlinear characteristic between input u and output $y(u, d)$ of a system affected by an additive output disturbance d (Figure 4), over the range of values $U = [0, 1]$ for u (input-output equalization).

The device is fed by $y(u, d)$ and produces output $y_{\mathcal{D}_\gamma}(u, d) = \gamma_1 y(u, d)^2 + \gamma_2 y(u, d) + \gamma_3$. The performance of the device with parameter $\gamma = (\gamma_1, \gamma_2, \gamma_3) \in \mathbb{R}^3$ is given by $\max_{u, d \in U \times D} J_\gamma(u, d)$, where $J_\gamma(u, d) = |y_{\mathcal{D}_\gamma}(u, d) - u|$

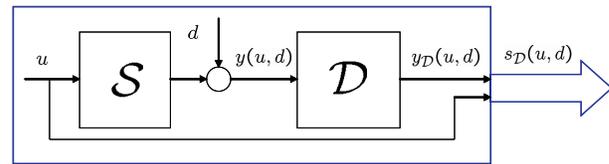


Fig. 4. Inverting a nonlinear characteristic through a device

and D is a (unknown) range of values for d . In words, this performance expresses the largest deviation off the perfect equalization line $y_{\mathcal{D}} = u$.

We chose $\epsilon = 0.1$, $\beta = 10^{-6}$, and according to (5) N was 205.

The scenario program (4) is in this case

$$\begin{aligned} \min_{\gamma, c \in \mathbb{R}^4} c \quad \text{subject to:} \quad (6) \\ |\gamma_1 y(u^{(i)}, d^{(i)})^2 + \gamma_2 y(u^{(i)}, d^{(i)}) + \gamma_3 - u^{(i)}| \leq c, \\ i = 1, 2, \dots, 205, \end{aligned}$$

where $u^{(1)}, u^{(2)}, \dots, u^{(205)}$ are random values for u independently extracted from U according to the uniform distribution P_u over $[0, 1]$, and $d^{(1)}, d^{(2)}, \dots, d^{(205)}$ are random values for d independently created by the environment during experimentation according to some (unknown) distribution P_d .

The 205 constraints in (6) can be evaluated by running 205 experiments on the system where the output samples $y^{(i)} = y(u^{(i)}, d^{(i)})$, $i = 1, 2, \dots, 205$, are collected together with $u^{(i)}$, $i = 1, 2, \dots, 205$. Figure 5 shows the outcomes of the experiments. Note that the collected output data present some dispersion due to the presence of the additive disturbance d .

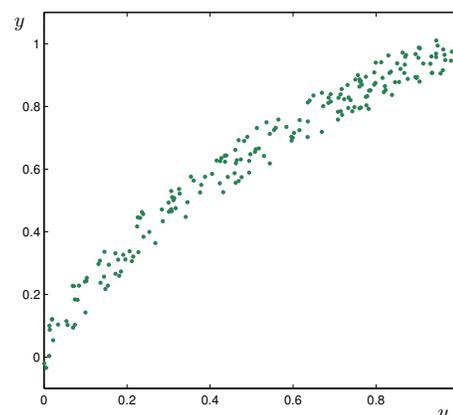


Fig. 5. Outcome of the experiments: samples of input u and output $y(u, d)$

By solving (6) we obtained $\gamma_{205}^* = (0.424, 0.650, -0.081)$ and $c_{205}^* = 0.108$.

c_{205}^* is the maximum equalization error for the extracted scenarios. In Figure 6, we plot the input and equalized output

pairs $(u^{(i)}, y_{D, \gamma_{205}^*}(u^{(i)}, d^{(i)}))$, $i = 1, 2, \dots, 205$, and the region $u \pm c_{205}^* := \{(u, y) : u - c_{205}^* \leq y \leq u + c_{205}^*, u \in U\}$. $u \pm c_{205}^*$ is the strip of minimum width centered around the perfect equalization line $y_D = u$ that contains all the 205 input and equalized output pairs.

In the light of the practical result at the end of the previous

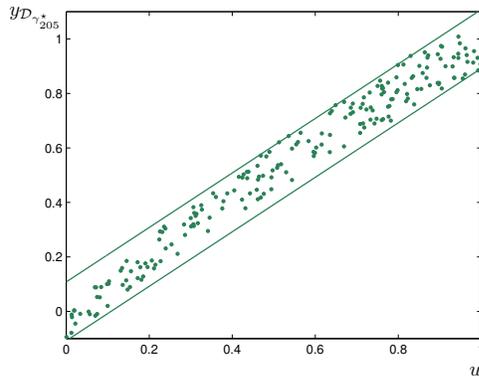


Fig. 6. Input u and equalized output $y_{D, \gamma_{205}^*}(u, d)$ for the extracted scenarios, and the region of equalization $u \pm c_{205}^*$

section, device γ_{205}^* carries a guarantee that the equalized output $y_{D, \gamma_{205}^*}(u, d)$ differs from u of at most $c_{205}^* = 0.108$ for all u 's and d 's except for a subset of probability $P = P_u \times P_d$ smaller than or equal to 0.1; moreover, the region of equalization $u \pm c_{205}^*$ is contained within $u \pm c^*$. This result holds irrespectively of D and P_d , which are unknown to the designer of the device.

The actual nonlinear characteristic and disturbance d used to generate the data in Figure 5 are shown in Figure 7 together with the designed device with parameter γ_{205}^* . In this example, the parameter of the device could have been designed so as to exactly invert the nonlinear characteristic. However, the obtained γ_{205}^* is different from such a choice, because the device aims at inverting the nonlinear characteristic between u and y while also reducing the effect of d on the reconstructed value for the input u .

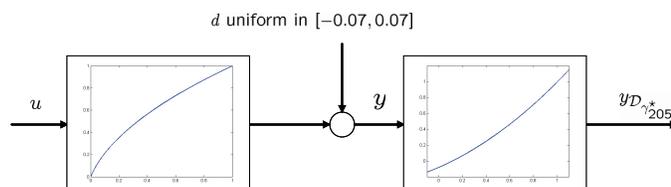


Fig. 7. Actual nonlinear characteristic and disturbance characteristics, along with the designed device

V. CONCLUSIONS

The main goal of this contribution is that of attracting the reader's attention to the fundamental issue of evaluating the experimental effort needed to perform adaptive design, and some answers have been provided in a specific worst-case context.

Many are the aspects that our discussion has left unanswered, and open to further investigation:

- it is not always the case that one experiment provides all the information needed to evaluate a constraint. In the disturbance compensator example, for instance, if either the system or the device are nonlinear it is not possible to swap them, and constraint evaluation calls for many experiments with virtually all possible devices in place. More generally, more experiments are needed when the input to the system depends on the device being designed.
- a perspective different from the worst-case approach can be used for adaptive design. For example, device quality could be assessed by its average performance, [10], [11], [12], rather than its worst-case performance over the set of operating conditions of interest, and the theory developed here does not apply to this case.

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